



$$Z_m = \frac{60}{\sqrt{\epsilon_r}} \ln \frac{a_2}{a_1}, \quad (1)$$

where  $Z_M$  is the characteristic impedance of the measurement system (50 Ω);  $\epsilon_r$  is the relative permittivity (in this case is equal to 1, air);  $a_1$ ,  $a_2$  are the radii of the coaxial line (flange).

The transition from the N-type connector to the opposite end of the flange has the linear shape for both parts of the flange, central and external one. This shape was chosen for better construction. The liner shape should be optimised for better impedance matching especially at frequencies over 1 GHz. The central flange conductor was constructed from brass. The rest of the flange was made from aluminium alloy. The flange has been tightened by the torque wrench after inserting the test composite by the same torque every time. This setup increases the accuracy of each measurement and also increases the repeatability during several measurements.

The measured scattering parameters of the flange itself are given in Fig.2. The measurements of scattering parameters were done according to [5]. The  $s_{11}$  and  $s_{22}$  are in the whole range of interest under -15 dB which refers about the good matching of both test ports with the measuring system. The insertion losses in both directions ( $s_{21}$  and  $s_{12}$ ) are in the whole frequency measuring range below 1 dB. This data refers about the accurate design of the whole coaxial flange. The flange itself will have insignificant influence on the total dynamic range of the whole measurement setup. The dynamic range will be mainly affected by the used measuring devices (by generator and spectral analyser). The measured scattering parameters refer about the accurate design of the coaxial flange.

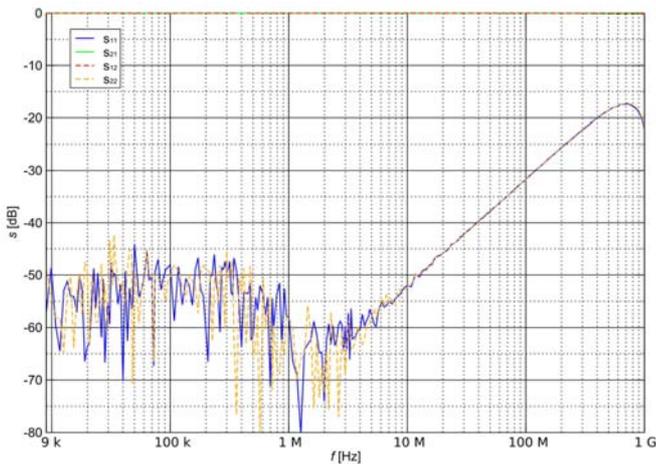


Fig.2 Measured scattering parameters of realised coaxial flange

### 3. VERIFICATION OF MEASUREMENT SETUP

Verification of the coaxial flange was done by measurement of the metal plate that forms the EMC chamber. The EMC chamber at the Department of Radio Electronics is made from exactly the same material. So at the first stage the shielding efficiency of the whole chamber was measured by the “classical test setup” with two antennas. Second measurement

data was produced by the coaxial flange. Obtained results are given in Fig.3. The measured data of the brass calibration ring was also added for comparison.

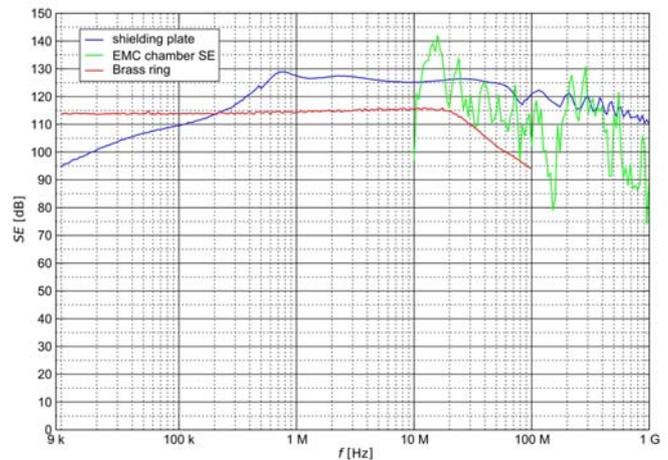


Fig.3 Performance comparison of the shielding measurement methods

Data presented in Fig.3 for the metal EMC chamber plate shows great agreement for both types of measurements. The results obtained in “classical” test setup are not so stable. The fluctuations are caused by the construction aspects of the EMC chamber. The chamber is not made from one full plate only, but it is bolted from several metal plates. Bolted connection reduces the whole shielding efficiency of the whole EMC chamber. This is the reason for the deviation of the total shielding efficiency. But the average progress corresponds very well with the data measured in the coaxial flange. In this flange the full plate without any defects or bolting was measured. Fig.3 shows great agreement between these two measurements. It could be established that the coaxial flange is appropriately designed and it is suitable for the intended shielding efficiency measurements.

### 4. SHIELDING EFFICIENCY MEASUREMENTS OF COMPOSITE MATERIALS

After the design and verification of the coaxial flange by the shielding measurements on the “known” material, the composite materials were measured. Firstly the test samples of concrete materials were produced. There were problems with the precise prefabrication of the test concrete rings. These rings have to be produced with high accuracy of their dimensions and the rings have to be firmly set into the flange. The real produced flange with the tested composite ring is depicted in Fig. 4. It is necessary to provide ideal connection between the centre conductor and the outer braid of the coaxial flange. In an ideal case, the composite material has to shorten the coaxial conductors. It means that the composite material has to have the lowest impedance possible for great shielding similar to the shielding plates. In another case the composite material has to have very high attenuation in longitudinal direction for attenuation of the RF energy inside the composite material.



Fig.4 Physical realisation of the coaxial flange

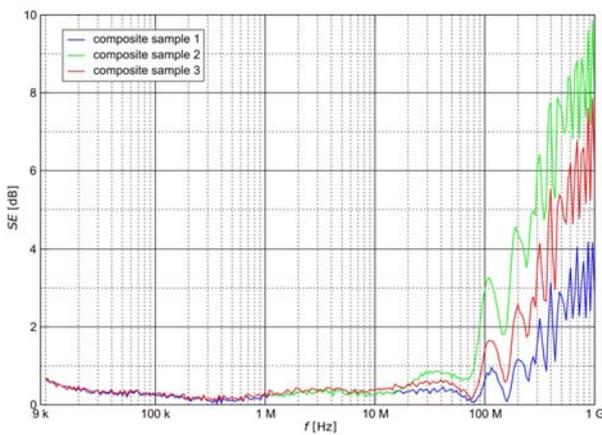


Fig.5 Shielding efficiency of the composite materials

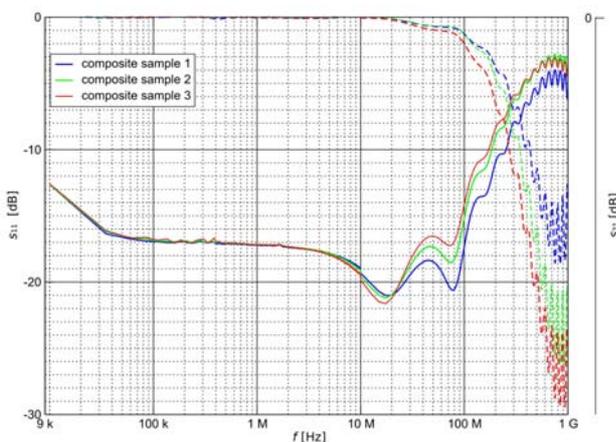


Fig.6 Scattering parameters of tested composite materials ( $s_{11}$  - full lines,  $s_{21}$  - dashed lines)

The shielding effectiveness measurements of the composite material were done on the test samples with about 8 mm thickness. The test sample is also shown in Fig.4. The measurements were focused mainly on the concrete with graphite during the development stage of these concrete-based materials as potential shielding materials. Some results are depicted in Fig.5. In this stage of the development, the shielding efficiency is not so high. The improvement of shielding efficiency is the main target of further development.

#### 5. DETERMINATION OF THE ELECTROMAGNETIC SHIELDING FACTORS FROM THE SCATTERING PARAMETERS

The total electromagnetic shielding is usually composed of and also influenced by several factors. These factors are mainly reflection loss, absorption loss and multiple path reflection losses. The knowledge of the quantity of each factor could be useful for the next development stage.

The realised coaxial flange is also suitable for that type of measurement. It is necessary to measure the scattering parameters  $s_{11}$  and  $s_{21}$ . Measured data for some composite samples are given in Fig.6. This data had to be divided into two parts and reflection and absorption losses could be determined. The multiple path reflection loss is not put into effect in this case. These losses are mainly significant in low frequency range.

The reflection  $SE_{RL}$  and absorption  $SE_{AL}$  losses could be calculated from measured scattering parameters  $s_{11}$  and  $s_{21}$  by the following formulas:

$$SE_{RL} = 10 \cdot \log \left( 1 - 10^{\frac{s_{11}}{10}} \right), \quad (2)$$

$$SE_{AL} = 10 \cdot \log \left( \frac{10^{\frac{s_{21}}{10}}}{1 - 10^{\frac{s_{11}}{10}}} \right). \quad (3)$$

Where  $s_{11}$  and  $s_{21}$  are measured scattering parameters by the coaxial test flange and  $SE_{RL}$  is determined reflection loss by the composite test flange. The  $SE_{AL}$  means the absorption loss in the composite material. The  $SE_{RL}$  coefficient represents reflections on both sides of the composite test sample. This fact is caused by the used measurement technique. The applied technique uses basic vector network analyser with peak detectors, so both reflections on the edges are added up into the one measured  $s_{11}$  parameter. The calculated reflection and absorption losses for some samples are depicted in Fig.7.

The data is shown for several composite materials with the thickness of 8 mm. By increasing the thickness two times, it becomes obvious that the shielding efficiency will not increase two times. In this case, the absorption loss will increase twofold only and reflection loss remains the same. The determined data flows make the increase in the conductivity of the tested samples necessary for better shielding. The increase in thickness of the material will not lead to the desired results.

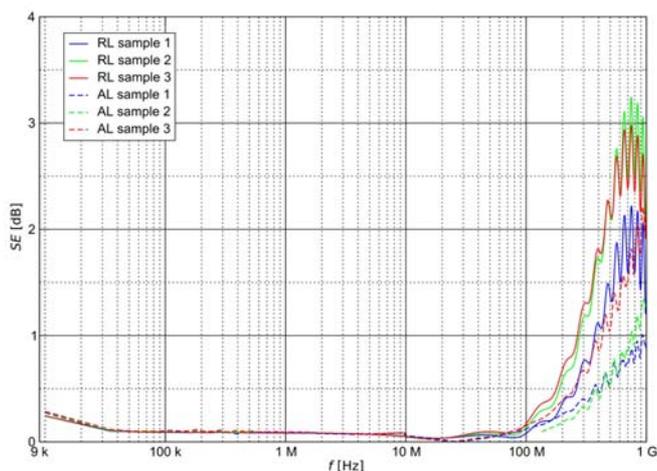


Fig.7 Calculated shielding efficiency components (reflection and absorption losses )

## 6. CONCLUSIONS

The modified test setup for the shielding efficiency measurements of the constructional materials like concrete or fabric was presented in the paper. The electric part of the shielding efficiency component should be measured by the shown setup and the measurement itself was done by the realised coaxial test flange. The properties of the flange were checked by several measurements. According to the measured data, it is possible to say that the coaxial test flange was optimally designed. Measured scattering parameters refer about the great matching of the coaxial test flange with the whole measuring system. The insertion loss caused by the flange is also negligible. There are not any dynamic range degradations. The presented test setup could provide the measurement with 140 dB dynamic range. Test flange was designed for the frequency range from 9 kHz up to 1 GHz. Indispensable optimization in the design will be required in the extension over 1 GHz. Linear shape transition should be replaced by any other shape.

The accuracy of the used test method with the flange was proven by the measurement of the same material two times. This material was used for the EMC chamber construction. The measurements were taken on the EMC chamber by the antennas and also by the coaxial flange. Measured results are in good agreement, which refers about the suitability of measurements with the test flange.

The scattering parameters of the composite materials were also measured. The reflection and absorption losses were determined according to this data (Fig. 7). From these results, it will be possible to establish the performance of the

composite material if their thickness increases. It is obvious that the increasing thickness will not lead to the desired performance of the e.g. 30 dB shielding efficiency. It is necessary to reduce the conductivity of the composite material, but this is not the aim of this article. The goal of this article was to introduce the measuring setup and discuss the parameters of the design flange and test setup. The high performance of the whole measurement setup was increased by the created control program in VEE Pro environment by which the measurements were controlled. The performance of the whole setup is on high level and the measurements produce accurate results.

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